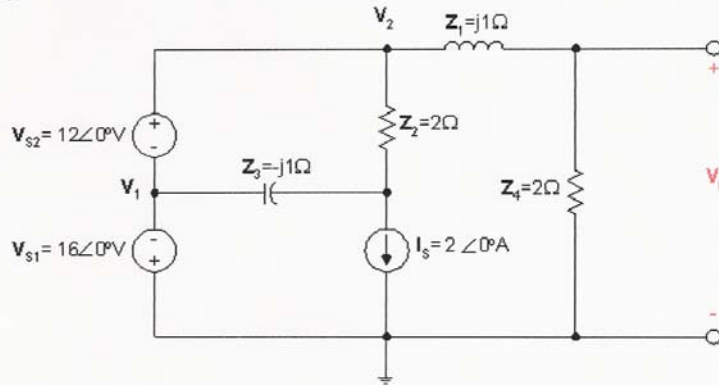


Problem 7.49

Solutions to #1

Problem
Hints
Solution Guidelines
Detailed Solution
Final Answer

- 1 Let us label the sources, node voltages, and impedances, and select a reference node as in Fig. P7.49a.



(Figure P7.49a)

- 2 First we note that the voltage source labeled V_{s1} is connected between node V_1 and the reference node, thus:

$$V_1 + V_{s1} = 0$$

$$V_1 = -16 \angle 0^\circ \text{V}$$

- 3 In addition, the voltage source labeled V_{s2} is connected between nodes V_1 and V_2 .

Therefore:

$$\begin{aligned} V_2 &= V_1 + V_{s2} = \\ &= -16 \angle 0^\circ + 12 \angle 0^\circ = -4 \angle 0^\circ \text{V} \end{aligned}$$

- 4 Knowing V_2 we can use voltage division to find V_0 .

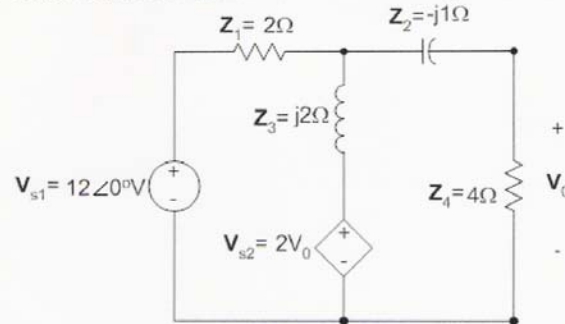
$$V_0 = V_2 \cdot \frac{Z_4}{Z_1 + Z_4} = -4 \left(\frac{2}{2 + j1} \right)$$

$$V_0 = 3.58 \angle 153.43^\circ \text{V}$$

Problem 7FE-3

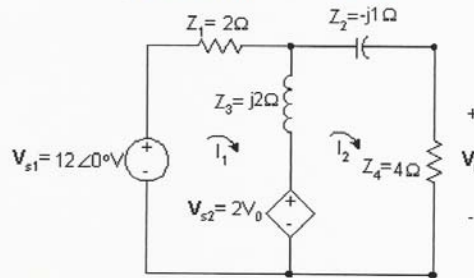
Problem Hints Solution Guidelines Detailed Solution Final Answer

1 Let us label the sources and the impedances as shown in Fig 7PEF-3a



(Figure 7PEF-3a)

2 Since the network has two meshes, we need two equations to determine all the currents. We will first assign one loop-current to each mesh as shown in Fig 7PFE-3b



3 The KVL equations for the meshes are:

$$-V_{s1} + I_1 Z_1 + (I_1 - I_2) Z_3 + V_{s2} = 0$$

$$I_2 Z_2 + I_2 Z_4 - V_{s2} + (I_2 - I_1) Z_3 = 0$$

4 Substituting the components values into the equations yields:

$$I_1 (2 + j2) + I_2 (-j2) = 12 - 2V_0$$

$$I_1 (-j2) + I_2 (4 + j) = 2V_0$$

5 $V_0 = I_2 \cdot Z_4$ by Ohm's law, thus:

5 $v_0 = I_2 \cdot Z_4$ by Ohm's law, thus:

$$I_1(2 + j2) + I_2(-j2) = 12 - 8I_2$$

$$I_1(-j2) + I_2(4 + j) = 8I_2$$

6 And solving for I_2 yields:

$$I_2 = \frac{12}{5 + j3} = 2.06 \angle -30.96^\circ \text{ A}$$

7 Knowing the loop current we can find V_0 using Ohm's law:

$$V_0 = I_2 \cdot Z_4 = 2.06 \angle -30.96^\circ \cdot 4 = 8.24 \angle -30.96^\circ \text{ V}$$

5 The KVL equations for the above network are:

$$-V_1 + R_1 I_1 + V_s = 0$$

$$-V_2 + R_2 I_2 = 0$$

6 Comparing the network to our standard circuit for mutually coupled inductors, we find that we must reverse the sign on I_1 and V_2 .

Therefore, the equations that relate V_1 and V_2 to I_1 and I_2 , in this case, are:

$$V_1 = -j\omega L_1 I_1 + j\omega M I_2$$

$$-V_2 = j\omega L_2 I_2 - j\omega M I_1$$

7 Combining the equations in (5) and (6) yields:

$$(j\omega L_1 + R_1) I_1 + (-j\omega M) I_2 + V_s = 0$$

$$(-j\omega M) I_1 + (R_2 + j\omega L_2) I_2 = 0$$

8 Substituting the components values into (7) we obtain :

$$(200 + j400) I_1 - j320 I_2 = -1$$

$$-j320 I_1 + (320 + j400) I_2 = 0$$

9 Solving the equations yields:

$$10 \quad I_1 = 2.46 \angle 143.1^\circ \text{ mA}$$

$$I_2 = 1.54 \angle -178.2^\circ \text{ mA}$$

11 Converting the currents back to time domain we obtain:

$$i_1(t) = 2.46 \cos(100t + 143.1^\circ) \text{ mA}$$

$$i_2(t) = 1.54 \cos(100t - 178.24^\circ) \text{ mA}$$

12 At $t = 2 \text{ ms}$, $100t = 0.2 \text{ rad}$ or 11.46° , and therefore,

$$i_1(t = 2 \text{ ms}) = 2.46 \cos(11.46^\circ + 143.1^\circ) = -2.22 \text{ mA}$$

$$i_2(t = 2 \text{ ms}) = 1.54 \cos(11.46^\circ - 178.24^\circ) = -1.5 \text{ mA}$$

13 Hence, the energy stored in the coupled inductors at $t = 2 \text{ ms}$ is

$$\begin{aligned} w(t)_{t=2\text{ms}} &= \frac{1}{2} L_1 [i_1(t=2\text{ms})]^2 + \frac{1}{2} L_2 [i_2(t=2\text{ms})]^2 - M i_1(t=2\text{ms}) i_2(t=2\text{ms}) \\ &= \frac{1}{2} \cdot 4 \cdot [(-2.22) \times 10^{-3}]^2 + \frac{1}{2} \cdot 4 \cdot [(-1.5) \times 10^{-3}]^2 \end{aligned}$$

<http://www.justask4u.com/csp/irwin/ESolutionDsp.csp?CSPCHD=00b00001000i3csv77a6003796045...> 9/23/2004

$$\begin{aligned} &= 3.2 \cdot (-2.22) \times 10^{-3} \cdot (-1.5) \times 10^{-3} \\ &= 3.7 \times 10^{-6} \text{ J} = 3.7 \mu\text{J} \end{aligned}$$

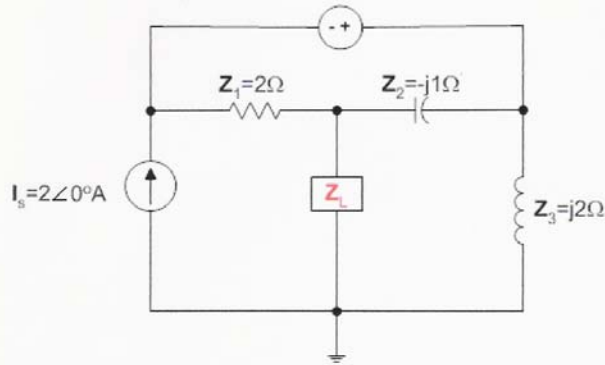
Problem 9FE-3

problem # 4

- Problem Hints Solution Guidelines Detailed Solution Final Answer

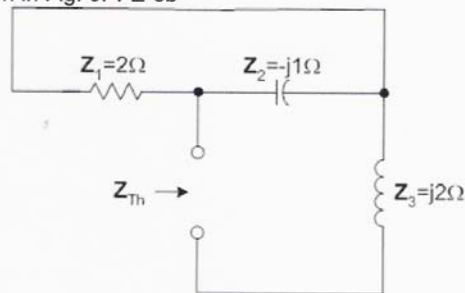
1 Let us label the impedances as shown in Fig. 9PFE-3a

$$V_s = 12 \angle 0^\circ \text{V}$$



(Figure 9PFE-3a)

- 2 To find the value of the load, Z_L , for maximum average power transfer, we simply need to find the complex conjugate of the Thévenin equivalent impedance.
- 3 The Thévenin equivalent impedance is obtained by looking into the open-circuit terminals with all sources made zero. In this case, we replace the voltage source with a short circuit and the current source with an open-circuit. The network is shown in Fig. 9PFE-3b



(Figure 9PFE-3b)

4 Combining the impedances yields:

$$Z_{Th} = Z_3 + \frac{Z_1 \cdot Z_2}{Z_1 + Z_2}$$

$$= j2 + \frac{2 \cdot (-j1)}{2 + (-j1)} = 0.4 + j1.2 \Omega$$

<http://www.justask4u.com/csp/irwin/ESolutionDsp.csp?CSPCHD=00h00001000i3csv77a6003378453...> 9/23/2004

5 Hence, for maximum average power transfer, the value of the load is the complex conjugate of the Thévenin equivalent impedance hence

$$Z_L = Z_{Th}^* = 0.4 - j1.2 \Omega$$